AMENDMENTS TO THE CLAIMS:

This listing of claims will replace all prior versions and listings of claims in the application:

- 1. (Currently Amended) A method for controlling a synchronous permanent magnet multiple-phase motor, the motor having multiple phases and having a rotor, the method comprising the steps of:
- [[-]] controlling [[the]] drive currents supplied to [[each]] the phases of the motor [[phase]] by turning [[it]] the drive currents off at a predetermined frequency[[,]];
- [[-]] measuring, at said predetermined frequency, [[the]] induced voltages of at least two of the [[motor]] phases of the motor, when the [[power]] drive currents in said at least two of the phases of the motor phase is are turned off, with a sensitivity allowing to obtain sufficient for obtaining significant voltages voltage values at a near-zero speed of the rotor;[[,]]
- [[-]] determining the rotor <u>a</u> position and/or <u>a</u> the rotor speed <u>of the rotor</u> from said measured induced voltages[[,]];
- [[-]] entering said determined rotor filtering the position and/or the saiddetermined rotor speed of the rotor [[into]] with a state filter which delivers to obtain a
 filtered rotor position and/or a filtered rotor speed[[,]]; and
- [[-]] adjusting the drive current as a function of said <u>currents according to the</u> filtered rotor position and/or the filtered rotor speed.

- 2. (Currently Amended) The method according to claim 1, wherein the state filter is arranged configured so as to take into account that when the speed of the [[motor]] rotor is very low, the position of the rotor [[can]] does not change substantially over a short period of time.
- 3. (Original) The method according to claim 2, wherein said state filter is a Kalman filter.
- 4. (Currently Amended) The method according to any of claims 1 to 3, wherein the measured position, θ , of the rotor is determined by the formula

$$\theta = \operatorname{arctg}\left(\frac{\mathsf{Ua}}{\mathsf{Ub}}\right)$$

where Ua is equal to the measured induced voltage in one of the phases of the motor [[phase]] and Ub is equal to $\frac{V2-V3}{\sqrt{3}}$, V2 and V3 being the measured induced voltages in two others of the ether two phases of the motor.

- 5. (Currently Amended) The method according to any of claims claim 1 [[to 4]], wherein the measured speed of the rotor is determined by computing [[the]] a square root of [[the]] a sum of squares of the measured induced voltages.
- 6. (Currently Amended) The method according to any of claims claim 2 [[to 5]], wherein the [[said]] state filter implements an algorithm such as

$$X = X_{-1} + (a * V * T + b * dP) \div 2,$$

where X is [[the]] an estimated position of the rotor at time t,

 X_{-1} is [[the]] an estimated position of the rotor at time t_{-1} ,

Xm is the measured position using back EMF voltages at time t_m with ($t_{\perp} \le tm \le t$),

V is [[the]] <u>a</u> measured speed <u>of the rotor</u> using back EMF voltages at time t_m , T is [[the]] <u>a</u> time duration between <u>2-successive measurements (t_1</u> and t[[)]], dP is the difference between Xm and X_1, <u>wherein Xm is a measured position</u> <u>of the rotor using back EMF voltages at time t_m with $t_1 \le t_m \le t$, and the [[such]] difference <u>dP is being however limited to $\pm (c * VT + d)$, and</u></u>

- a, b, c and d are coefficients which depend on <u>characteristics of</u> the motor-characteristics.
- 7. (Currently Amended) An electronic device for controlling a synchronous permanent magnet motor [[(1)]] with at least one phase, a coil, a rotor, and a motor driver [[(2)]], the electronic device comprising:

detection means (3), which are connected to the <u>at least one phase</u> phases (A, B, C) of the motor and deliver for generating signals that represent induced voltages of the at least one phase of the motor phases, said detection means having a high enough gain to provide significant output such that the signals representing the induced voltages are significant even if [[the]] a speed of the rotor is near-zero[[,]]; and

a control circuit [[(4)]] connected to said detection means and to the motor driver (2), which supplies for supplying driving currents to the motor, said control circuit comprising means for generating signals representing a computing the position and/or

[[the]] <u>a</u> speed of the rotor from the output signals provided by said detection means representing the induced voltages.

8. (Currently Amended) The electronic device according to claim 7, wherein the motor includes at least two phases, and wherein said detection means comprise comprises, for each of the at least two phases phase of the motor,:

a differential amplifier (31, 32, 33) the <u>having</u> inputs of which are connected to two of the at least two different phases of the motor; and

an analog-to-digital converter (34, 35, 36) to convert the for converting an analog signal outputted by said differential amplifier into a digital signal, which is applied and providing said digital signal to said control circuit.

- 9. (Currently Amended) The electronic device according to claim 7 or claim 8, wherein the control circuit further comprises a state filter for filtering the signals representing the position and/or the speed of the rotor meter determined from the output signals of said detection means.
- 10. (Currently Amended) The electronic device according to claim 9, wherein said state filter is a Kalman filter.
- 11. (New) The method according to claim 4, wherein the speed of the rotor is determined by computing a square root of a sum of squares of the measured induced voltages.

12. (New) The method according to claim 4, wherein the state filter implements an algorithm such as

$$X = X_{-1} + (a * V * T + b * dP) \div 2$$
,

where X is an estimated position of the rotor at time t,

 X_{-1} is an estimated position of the rotor at time t_{-1} ,

V is a measured speed of the rotor using back EMF voltages at time t_m,

T is a time duration between t₋₁ and t,

dP is the difference between Xm and X_{-1} , wherein Xm is a measured position of the rotor using back EMF voltages at time t_m with $t_{-1} \le tm \le t$, and the difference dP is limited to $\pm (c * VT + d)$, and

a, b, c and d are coefficients which depend on characteristics of the motor.

13. (New) The method according to claim 5, wherein the state filter implements an algorithm such as

$$X = X_{-1} + (a * V * T + b * dP) \div 2$$
,

where X is an estimated position of the rotor at time t,

 X_{-1} is an estimated position of the rotor at time t_1 ,

V is a measured speed of the rotor using back EMF voltages at time t_{m} ,

T is a time duration between t₋₁ and t,

dP is the difference between Xm and X_{-1} , wherein Xm is a measured position of the rotor using back EMF voltages at time t_m with $t_{-1} \le t_m \le t$, and the difference dP is limited to $\pm (c * VT + d)$, and

a, b, c and d are coefficients which depend on characteristics of the motor.